



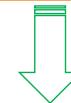
Stabilization control of inverted pendulum



FL09-7-24
Hironori Itoh

Introduction

Invert pendulum
e-nuvo WHEEL(ZMP INC.)
had no controller with d-SPACE.



- Derive equation of motion.
- Control with d-SPACE.
- Apply some controller.



Outline

- ✓ Adapt to d-SPACE
 - Equation of motion
 - Calibration of potentiometer
 - LQ controller
 - Filter
- LQI Controller
- PID Controller
- Conclusion



Equation of motion(1)

Cart position

$$(x_c, y_c) = (l \sin \theta + r_i \phi, -l \cos \theta)$$

Pendulum position

$$(x_p, y_p) = (r t \phi, 0)$$

Lagrange equation of motion

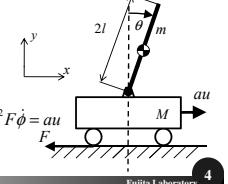
$$\frac{d}{dt} \left(\frac{\partial L}{\partial \dot{q}} \right) - \frac{\partial L}{\partial q} = 0$$

Result of Lagrange equation of motion

$$(ml^2 + Jp)\ddot{\theta} + c\dot{\theta} - gml \sin \theta + mlr_i \ddot{\phi} = 0$$

$$mlr_i \ddot{\theta} \cos \theta - mlr_i \dot{\theta}^2 \sin \theta + [(M+m)r_i^2 + Jt + i^2 Jm]\ddot{\phi} + r_i^2 F\dot{\phi} = au$$

m : Mass of the pendulum
 M : Mass of the cart
 r_{at} : Reduction ratio of the gear
 Jp : Moment of inertia of the pendulum
 Jt : Moment of inertia of the cart
 Jm : Moment of inertia of the motor rotor
 l : Length between the pendulum axle and the gravity center of the cart
 r_i : Radius of the wheel
 F : Viscous Friction of the system
 c : Friction of the wheel axle
 Kt : Torque constant of the motor
 a :



Equation of motion(2)

Linearize the equation of motion with $\cos \theta \approx 1, \sin \theta \approx \theta$

$$\ddot{\theta} = \frac{(M+m)Jt + l^2 Jm/r_i^2 - c\dot{\theta} + mg(l\theta) + mlr_i F\dot{\phi} - mlr_i au/r_i^2}{(ml^2 + Jp)(M+m) + Jt/r_i^2 + i^2 Jm/r_i^2 - (ml)^2}$$

$$\ddot{\phi} = \frac{ml/r_i(c\dot{\theta} - mg\theta) - F(ml^2 + Jp)\dot{\phi} + (ml^2 + Jp)au/r_i^2}{(ml^2 + Jp)(M+m) + Jt/r_i^2 + i^2 Jm/r_i^2 - (ml)^2}$$

$$= \alpha = \beta$$

Give the state equation with state $x = (\theta, \phi, \dot{\theta}, \dot{\phi})^\top$

$$\begin{bmatrix} \frac{dx}{dt} \\ y \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ \frac{cmlg}{\beta} & 0 & -\frac{\alpha c}{\beta} & \frac{mlr_i}{\beta} \\ 0 & 0 & 0 & \frac{1}{\beta} \\ \frac{m^2 l^2 g}{r_i \beta} & 0 & \frac{mlc}{r_i \beta} & -\frac{F(ml^2 + Jp)}{\beta} \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ \frac{mlr_i a}{r_i^2 \beta} \\ \frac{(ml^2 + Jp)a}{r_i^2 \beta} \end{bmatrix} u$$

$$\begin{bmatrix} dx \\ dt \\ y \end{bmatrix} = Ax + Bu$$

$$y = Cx + D$$

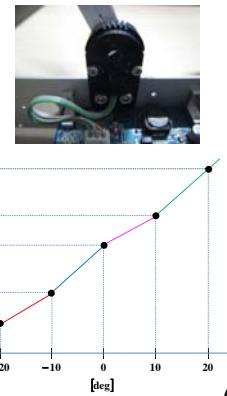


Calibration of potentiometer(1)

Interpolation line

$$\theta_{deg} = \begin{cases} \frac{I_{-10} - I_{-20}}{10} I - 20 + \frac{I_{-10} - I_{-20}}{10} I_{-20} & (I \leq I_{-10}) \\ \frac{I_0 - I_{-10}}{10} I - 10 + \frac{I_0 - I_{-10}}{10} I_{-10} & (I_{-10} \leq I \leq I_0) \\ \frac{I_{10} - I_0}{10} I + 0 + \frac{I_{10} - I_0}{10} I_{10} & (I_0 \leq I \leq I_{10}) \\ \frac{I_{20} - I_{10}}{10} I + 10 + \frac{I_{20} - I_{10}}{10} I_{20} & (I_{10} \leq I) \end{cases}$$

$$\theta_{rad} = \frac{2\pi}{360} \theta_{deg}$$

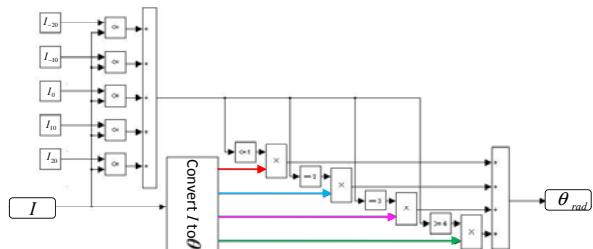




Calibration of potentiometer(2)

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Model of Calibration



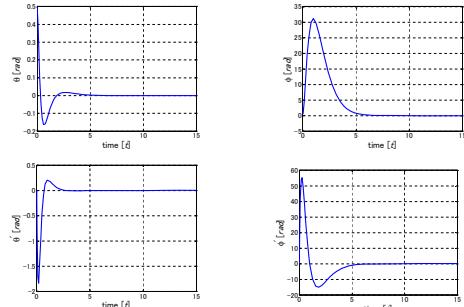
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LQ controller (2)

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$$x_0 = [0.5 \quad 0.5 \quad 0 \quad 0]^T$$

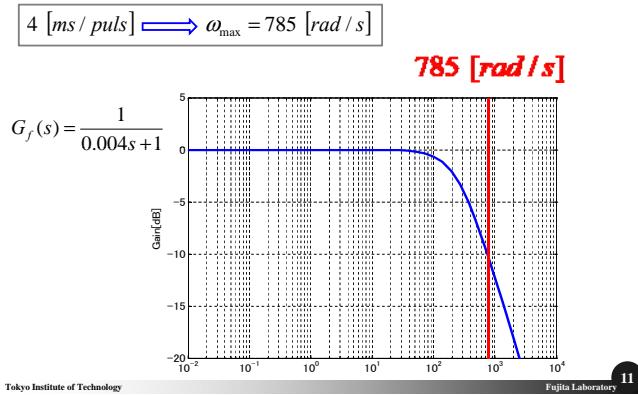
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Filter (1)

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LQ Controller (1)

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LQ controller

Quadratic evaluation function

$$J = \int_0^{\infty} (x^T Q x + u^T R u) dt$$

When J is minimum

$$u = -R^{-1} B^T P x = -K x$$

, where P is positive definite solution of Riccati equation

$$P A + A^T P - P B R^{-1} B^T P + Q = 0 \quad Q \text{ (weighting matrix)} \in \Re^{4 \times 4} \\ R \text{ (weighting matrix)} \in \Re$$

Ex.

$$Q = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, \quad R = 500 \quad \Rightarrow \quad K = [-31.7 \quad -5.7 \quad -0.0045 \quad -0.086]$$

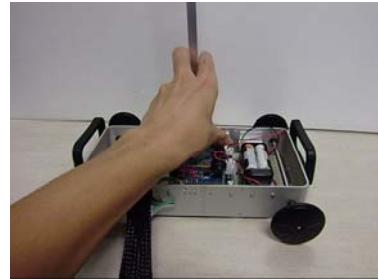
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Filter (2)

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Transfer function of a filter

$$X(s) = \frac{1}{Ts + 1}$$

Z Transformation

$$X(z) \equiv (1 - z^{-1}) Z\left[\frac{X(s)}{s}\right]$$



$$\therefore X(z) = \frac{1 - \exp(-\tau/T)}{z - \exp(-\tau/T)}$$

Ex.) $\tau = 0.004 \text{ [s]}$

$$\textcircled{1} T = 0.004 \quad X(s) = \frac{1}{0.0025s + 1} \\ \Rightarrow X(z) = \frac{0.2212}{z - 0.7788}$$

② $T = 0.01$

$$X(s) = \frac{1}{0.01s + 1} \\ \Rightarrow X(z) = \frac{0.09516}{z - 0.9048}$$

③ $T = 0.1$

$$\Rightarrow X(z) = \frac{0.00995}{z - 0.99}$$

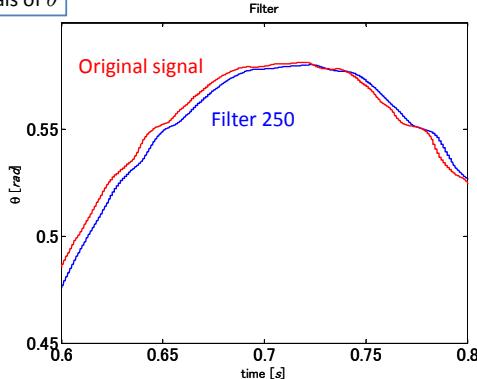
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Filter (3)

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Signals of θ 

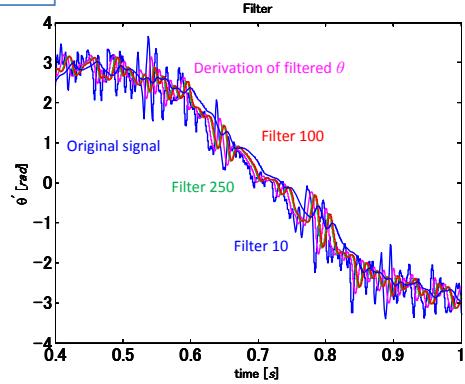
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Filter (4)

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Signals of $\dot{\theta}$ 

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Outline

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- Adapt to d-SPACE
- ✓ LQI Controller
- PID Controller
- Conclusion

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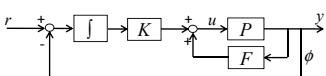


Servo system

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Servo integrator

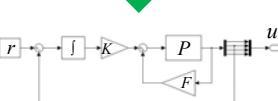
$$\frac{d\eta}{dt} = r - \phi \\ = r - [0 \ 1 \ 0 \ 0] x = r - C_\phi x$$



Augmented system

$$\frac{d}{dt} \begin{bmatrix} x \\ \eta \end{bmatrix} = \begin{bmatrix} A & 0 \\ -C_\phi & 0 \end{bmatrix} \begin{bmatrix} x \\ \eta \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u + \begin{bmatrix} 0 \\ I \end{bmatrix} r$$

$$y = [C \ 0] \begin{bmatrix} x \\ \eta \end{bmatrix}$$



Riccati Equation

$$PA_d + A_d^T P - PB_d R^{-1} B_d^T P + Q = 0$$

$$[F \ K] = -R^{-1} B_d^T P$$

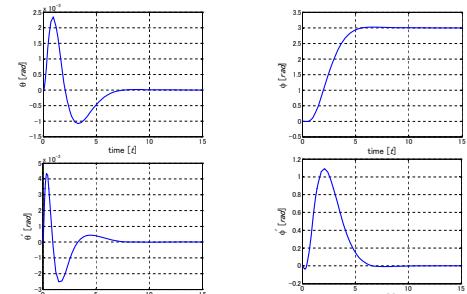
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Simulation

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$$Q = I \quad x_0 = [0 \ 0 \ 0 \ 0]^T \\ R = 500 \quad r = [0 \ 3 \ 0 \ 0]^T$$

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$r = \text{const}$

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Outline

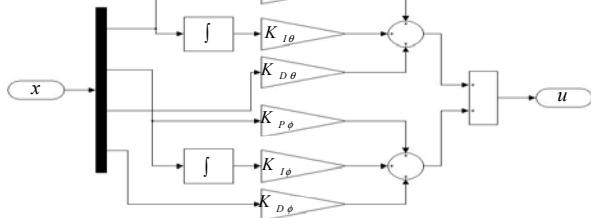
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- Adapt to d-SPACE
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PID controller

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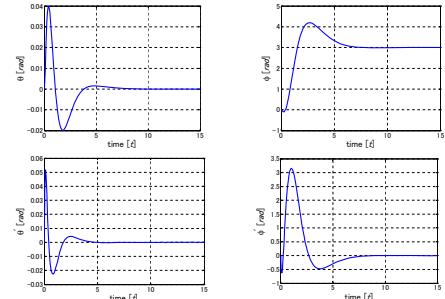
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Simulation

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$$x_0 = [0 \ 0 \ 0 \ 0]^T \quad [K_{P\theta} \ K_{I\theta} \ K_{D\theta}] = [-36.1 \ 0 \ -6.47]$$

$$r = [0 \ 3 \ 0 \ 0]^T \quad [K_{P\phi} \ K_{I\phi} \ K_{D\phi}] = [-0.11 \ -0.045 \ -0.13]$$

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$r = \text{const}$

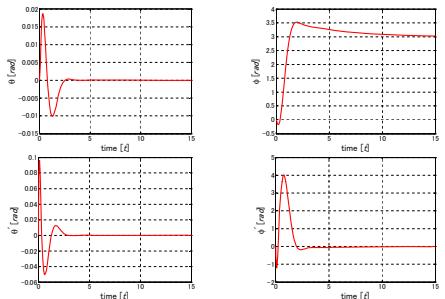
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Tuned PID

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$$x_0 = [0 \ 0 \ 0 \ 0]^T \quad [K_{P\theta} \ K_{I\theta} \ K_{D\theta}] = [-44.3 \ 0 \ -8.0]$$

$$r = [0 \ 3 \ 0 \ 0]^T \quad [K_{P\phi} \ K_{I\phi} \ K_{D\phi}] = [-0.24 \ -0.045 \ -0.21]$$

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$$r = 3 \sin \omega t$$

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Outline

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- Adapt to d-SPACE
- LQI Controller
- PID Controller

✓ Conclusion

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Conclusion

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Conclusion

- Controlled inverted pendulum with d-SPACE
- Applied some basic controllers

Future works

- Analyze with transfer function
- Apply another controller

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